

Pulse Generator

Model 8020

# INSTRUCTION MANUAL

BERKEL

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#### PARTS LIST AND SCHEMATICS

Berkeley Nucleonics Corporation warrants all instruments, including component parts, to be free from defects in material and workmanship, under normal use and service for a period of one year. If repairs are required during the warranty period, contact the factory for component replacement or shipping instructions. Include serial number of the instrument. This warranty is void if the unit is repaired or altered by others than those authorized by the Berkeley Nucleonics Corporation. 



SPECIFICATIONS

- REPETITION RATE: a) 0.5 Hz-10 Hz, continuously variable, 3 kHz-125 MHz continuously variable.
  - b) Ext Trigger, 0-125 MHz.
  - c) Single Cycle.
- DELAY: 0 to 100 µsec, continuously variable.
- WIDTH: 3 nsec to 100 µsec, continuously variable.
- JITTER: Rep rate, delay or width less than 50 psec or 0.1%, whichever is greater.
- DOUBLE PULSE: 6 nsec min. separation. Pulse spacing set by delay controls.
- **RESOLUTION OF FINE CONTROLS: Less than 0.4%.**
- TEMPERATURE COEFFICIENT OF FREQUENCY, DELAY OR WIDTH: Less than 0.1%/°C.
- DUTY FACTOR: Greater than 50%.
- OUTPUT PULSES: Two parallel output connectors providing greater than --32 mA. When terminated in 50 Ω, there are two --0.8 V pulses. (Standard NIM fast logic level.)
- RISE TIME: 1 nsec.
- FALL TIME: 1.3 nsec.
- AMPLITUDE ADJUSTMENT: 10:1 range (from -32 mA to -3.2 mA), continuously variable.
- OUTPUT PULSE ABERRATIONS: Baseline or pulse top, less than 5%.
- TRIGGER OUT: -0.8 V, 50  $\Omega$ , 1 nsec rise time. (Two outputs on front and rear panel.)
- EXTERNAL TRIGGER: ---0.6 V, 50 Ω. (Rear panel.)
- EXT GATE: NIM logic. (--0.6 V to gate on, at rear panel.) Synchronous. Rear panel slide switch provides gated or ungated operation.

AMBIENT TEMPERATURE: 55°C max.

PROTECTION: Open and short circuit proof.

- POWER REQUIREMENTS: +24 V, 120 mA, +12 V, 120 mA, -24 V, 135 mA, -12 V, 135 mA.
- MECHANICAL: Single width AEC module, 1.35" wide x 8.70" high in accordance with TID-20893 (Rev. 2).

WEIGHT: 21/2 lbs., net, 7 lbs. shipping.

### OPERATING INSTRUCTIONS

#### 2.1 INTRODUCTION

The Model 8020 pulse generator is a one nanosecond rise time, 125 MHz pulse generator specifically designed for use with high energy physics instrumentation. However, its broad range of functions is applicable in other areas. It provides two parallel outputs of -16 mA which convert to two -0.8 V pulses terminated into 50 ohms.\*

The Model 8020 conforms to the mechanical and electrical specifications for NIM (nuclear instrument modules) set up by AEC report TID-20893, Rev. 3. In order to power this instrument it is necessary to provide +24 V and +12 V from an external supply. Berkeley Nucleonics manufactures portable power supplies, the Model AP-1 and Model AP-2, which are designed to provide the necessary power.

## 2.2 FUNCTION OF CONTROLS AND CONNECTORS

FREQUENCY: A twelve - position switch and 25-turn trimming potentiometer provide the range desired. The internal repetition rate is adjustable from 3 kHz to 125 MHz. Below 3 kHz, a range of frequencies from 0.5 to 10 Hz is available by setting the FREQUENCY switch to 10 Hz. In addition a S.C. (single cycle) pushbutton and EXT (external trigger) operation is provided.

DELAY: A ten-position switch and a 25-turn trimming potentiometer provides a continuous delay between PULSE OUT and TRIG OUT from  $0 - 100 \mu sec$ .

WIDTH: A ten-position switch and 25-turn trimming potentiometer provides a continuous pulse width from 3 nsec to 100 µsec.

S.P./D.P. (Single Pulse / Double Pulse): A front panel toggle provid is either a single or double output pulse with each Trigger Out pulse. When this control is in the S.P. (single pulse) position, one output pulse appears with each Trigger Out pulse. When this toggle is in the D.P. (Double Pulse) position, two output pulses appear with each Trigger Out pulse. The first pulse is time coincident with the trigger out and second pulse is separated from the first by the setting of the delay controls.

\*This logic level has been adopted by the U.S. AEC committee on nuclear instrument modules. See Standard Nuclear Instrument Modules, Report TID-20893 Rev. 3, available from the U.S. Government Printing Office, Washington, D.C. 20402 for 40 cents. AMPL. (Amplitude): A front panel mounted trimmer adjusts the output amplitudes over a 10:1 range. Normally it is set at the maximum to provide -16 mA output into each of two 50 ohms output terminations (or 0.8 V at each output when terminated in 50 ohms.)

GATE - ON/OFF: A rear panel slide switch provides gated or ungated operation. When it is in the GATE-OFF position a continuous pulse train appears at the output. When it is in the GATE-ON position a pulse burst will appear at the output connector only during the time that a -0.6 V gate signal appears at the rear panel gate connector.

TRIG OUT (Trigger Out): A connector on both the front and rear panel provides synchronizing pulses in time relation to the output pulse as set by the Delay controls. The Trigger Out pulses are a square wave train whose period is the same as the Frequency clock. THE TRIGGER OUTPUT CABLE MUST BE TERMINATED INTO 50 OHMS FOR PROPER OPERATION. The unused output need not be terminated.

PULSE OUT: Two parallel connectors on the front panel provide identical output pulses. When the AMPL (Amplitude) trimmer is set at maximum, -32 mA is available. The -32 mA divides into two -16 mA outputs when each connector is terminated in 50 ohms. If only one connector is terminated, the output amplitude will be -1.6 V across the 50 ohms termination and the rise time will be slower than 1 nanosecond. IT IS NECESSARY TO TERMINATE BOTH OUTPUT CONNECTORS IN 50 OHMS TO OBTAIN 1 NS RISE TIME AND -0.8 V PULSES.

EXT TRIG (External Trigger): A rear panel connector is provided to accept external synchronizing pulses. The FREQUENCY switch must be in the EXT position for this mode of operation. The input impedance is 50 ohms, and a -0.6 V amplitude pulse with a rise time less than 0.2 µsec is required.

GATE: A rear panel connector is provided to receive the external gate pulse. The input impedance is 50 ohms and -0.6 V is required to gate on the pulses. See GATE-ON/OFF above.

#### 2.3 OPERATIONAL CHECK POINTS

If there are any difficulties in obtaining desired operation from the Model 8020, the following check list will be useful.

#### NO OUTPUT PULSE:

- Is the rear panel GATE ON/OFF in the proper position? (Set at GATE-OFF for continuous pulse train).
- Is the front panel FREQUENCY switch set correctly?

#### POOR OUTPUT PULSE SHAPE:

- Are both output connectors terminated in 50 ohms (even if one is not used)?
- Is the output cable 50 ohms characteristic impedance?

#### ERRATIC TRIGGERING OF OSCILLOSCOPE:

- Is the Trigger Out cable terminated in 50 ohms?
- Does the delay or width duty factor exceed 50%?

#### ERRATIC EXTERNAL TRIGGER OPERATION:

- Is the external trigger at least -0.6 V in amplitude when connected?
  - Is the front panel FREQUENCY switch set at EXT?

### DIFFICULTY IN EXTERNAL TRIGGERING AT 125MHz REP RATE:

Is the Fine Frequency Control set at maximum counter-clockwise position?

#### DIFFICULTY IN OBTAINING 125MHz REPRATE:

Are the Delay and Width Controls set near minimum so that 50% duty factor is not exceeded?

EXCESSIVE JITTER: Check power supply regulation.

INSUFFICENT AMPLITUDE: Check setting of AMPL. control.

### CIRCUIT DESCRIPTION

This section of the manual describes the circuits used in the Model 8020. Section 3.1 describes the block diagram and reference is made to Fig. 1. Sections 3.2-3.11 describe the detailed circuits and reference is made to the schematics 8020-1, 8020-2 and 8020-3 at the back of the manual. See Sect. 3.12 for service.

#### 3.1 BLOCK DIAGRAM

A block diagram of the Model' 8020 is shown in Fig. 1. The clock pulses are generated by either the 3 kHz-125 MHz Multivibrator or the 0.5 Hz-10 Hz Multivibrator. The output of the 3 kHz-125 MHz Multivibrator triggers the Delay One-Shot and also triggers the Trigger Out circuit. The output of 0.5 Hz-10 Hz Multivibrator triggers the External Trigger circuit which in turn triggers the Delay One-Shot and Trigger Out circuit. When either the 0.5 Hz-10 Hz er 3 kHz-125 MHz Multivibrator is operating the other is inhibited.

The External Gate is connected directly to the 3 kHz-125 MHz Multivibrator. When the rear panel gate switch is in the "ON" position, the multivibrator is inhibited until a gating signal appears at the External Gate connector. The multivibrator then generates clock pulses for a period of time equal to the gate pulse width.

The External Trigger circuitry accepts external synchronizing pulses and shapes them to provide a trigger for the Delay One-Shot and the Trigger Out circuit. The 0.5 Hz-10 Hz and 3 kHz-125 MHz Multivibrators are inhibited during External Trigger operation.

The Delay One-Shot generates a pulse whose width is adjustable by the front panel Delay controls. The output of this one-shot passes into the Single/Double Pulse circuitry. If this circuit is set in the Single Pulse mode a trigger pulse is generated coincident with the trailing edge of the delay pulse. If the circuit is set in the Double Pulse mode, an additional trigger pulse appears which is coincident with the leading edge of the delay pulse.

The output of the Single/Double circuitry triggers the Width One-Shot. This one-shot generates a pulse whose width is adjustable by front panel Width controls. The Output Pulse Shaper generates the proper pulse shape to drive the output circuitry. The output circuitry is a current switch which provides a current of -36 mA into the output terminating resistances.

#### 3.2 3 kHz-125 MHz MULTIVIBRATOR

Refer to Schematic 8020-1. Transistors Qll-Ql8 function as a free-running multivibrator from 3 kHz-125 MHz. The feedback loop for regeneration is from the collector of Q11 through R36 and C18 to the base of Q12. The emitter of Q12 is capacitively coupled to the emitter of Q11. The value of the coupling capacitor C33-C41 sets the frequency range. The base of Q11 is coupled to the collector Q12 through C17 and R35, which completes the feedback loop.

Transistors Ql4 and Ql5 are constant current sources for the emitters of Ql1 and Ql2. Transistors Ql6 inhibits the multivibrator when the Frequency switch is in the 10 Hz or External/ Single Cycle position. Transistor Ql3 is connected as an emitter follower to provide +9.3 V to the collectors of Ql1 and Ql2.

The fine frequency setting of the multivibrator is controlled by adjusting the clamping levels of the collector waveform via Dl2 and Dl3. This clamping level is set by the Fine trimmer R58 via Ql7, Dl4, and Ql8. The function of Dl4 and Ql8 is to compensate for the temperature coefficient of Ql7, Dl2, and Dl3.

The output of the multivibrator appears across R33 and R34 and is derived from the collectors of Q11 and Q12 through R45 and R46. These pulses are referred to as clock pulses and drive the bases of Q24 and Q25.

#### 3.3 0.5 Hz-10 Hz MULTIVIBRATOR

Refer to Schematic 8020-1. The low frequency multivibrator of 0.5 Hz-10 Hz is provided by transistors Q54-Q58. The feedback loop is from the emitter of Q54 through C28 to the emitter of Q55, and from the collector of Q55 to the base of Q54. The timing capacitor is C28 which receives its charging current from the collector of Q56. The amount of charging current is set by the Fine frequency trimmer, R58, which controls the current through Q57 and Q56. The output of the multivibrator at the collector of Q54 drives the base of Q58. The collector of Q58 provides a trigger to the base of Ql, which is the input to the External Trigger The description of the operation of circuit. this circuit is given next in Section 3.4.

#### 3.4 EXTERNAL TRIGGER

Refer to Schematic 8020-1. Transistors Ql-Q6 function to shape an external trigger to provide a clock pulse in place of the free-running multivibrators. Transistors Ql and Q2 are in a Schmitt Trigger configuration. Transistor Ql is normally conducting and Q2 is normally nonconducting. This state is reversed when an External Trigger pulse is connected to the base of Ql. Transistor Q3 sets the bias on the base of Q2. Transistors Q4 and Q5 function as a current switch to increase the gain of the Schmitt Trigger. The current output of Q4 and Q5 appears across R33 and R34 to produce pulses to trigger the bases of Q24 and Q25. Transistor Q6 functions as an inhibit gate for the External Trigger circuit when the Frequency switch is in any position except External or 10 Hz. Transistor Q53 inhibits the external trigger circuit when the gating circuit is operating.

#### 3.5 EXTERNAL GATE

Refer to Schematic 8020-1. Transistors Q7-Q10 function to gate off the 3 kHz-125 MHz multivibrator during a period of time an external gate signal is present. Transistors Q7 and Q8 function as a Schmitt Trigger. During the gated mode of operation, Q7 is normally conducting and Q8 is non-conducting. A negative gate at the base of Q7 reverses this state. Transistor Q9 operates as an emitter-follower to provide the bias for the base of Q8. The circuit functions in the following manner:

When the GATE/ON-OFF switch is in the OFF position, the voltage at the base of Q9 is  $\pm 1.4$  V. The voltage at the base of Q8 is  $\pm 0.85$  V. The collector of Q8 is  $\pm 2.6$  V and the emitter of Q10 is  $\pm 2.0$  V. Diode D7 is back-biased and the multivibrator is unaffected by a gate signal.

When the GATE/ON-OFF switch is in the ON position, the voltage at the base of Q9 is  $\pm 0.2$  V which produces a voltage of -0.4 V at the base of Q8. The collector of Q8 is at  $\pm 6.4$  V and the emitter of Q10 is at  $\pm 5.8$  V. This voltage, through diode D7, clamps the collector of Q14 and the multivibrator is inhibited. When a negative gate pulse appears at the base of Q7, these conditions are reversed and the emitter of Q10 shifts to  $\pm 2.0$  V. Diode D7 is backbiased and the multivibrator functions for a period of time equal to the gate pulse width.

#### 3.6 DELAY LINE DRIVER & TRIGGER OUT

Refer to Schematic 8020-2. The clock pulses from the frequency multivibrators or external trigger circuit provide complementary signals which drive the bases of Q24 and Q25. Transistors Q24 and Q25 are in a current-switch configuration and their purpose is to both shape the pulses for the Trig Out (Trigger Out) circuits and to drive the Delay One-Shot.

The pulses at the collector of Q24 are delayed for 8 ns through DL-land then trigger the bases of Q19 and Q21. Transistors Q19, Q20, Q21, and Q22 function as current switches to provide Trigger Out pulses on both of the front and rear panels. Transistor Q23 is an emitter-follower to provide -2.7 V to the bases of Q19 and Q21. The bases of Q20 and Q22 are at -2.5 V which is provided by the voltage drop across D19. The output pulses at the collectors of Q20 and Q22 appear at the two TRIG OUT connectors. Diodes D17 and D18 at the collector of Q20 (and diodes D20 and D21 at the collector of Q22) provide a return path to ground if the TRIG OUT connector is not terminated.

The pulses at the collector of Q25 function to trigger the Delay One-Shot. At the collector of Q25 is a shorted delay line DL-2, with a double transit time of 4 nsec. This provides a clipped pulse of approximately 3 ns width at half-height. This pulse is passed through emitter-follower Q26 and then triggers the Delay One-Shot, Q27-Q31.

#### 3.7 DELAY ONE-SHOT

Refer to Schematic 8020-2. Transistors Q27, Q28 and Q29 form the regenerative loop for the Delay One - Shot. Before a trigger pulse arrives, the base of Q27 is at.31 Vand the base of Q28 is at +.65V. Transistor Q27 is nonconducting and Q28 is in conduction. A +0.6V trigger pulse at the base of Q27 starts to reverse this state and the collector of Q27 goes negative. This excursion passes through emitter-follower Q29, through timing capacitors C64-C72, and then to the base of Q28. This pulse cuts off Q28 which causes Q27 to go into full conduction. A quasi-stable state now exists until the timing capacitor charges up through When the base of Q28 nearly R87 and R82. reaches 0 V, Q28 starts to conduct and Q27 starts to cut off. The transistors, by regenerative action, then switch back to their inital Transistor Q29 is a low impedance states. source to rapidly discharge the timing capacitor to its initial state.

The coarse timing period of the quasi-stable state is provided by front panel selection of one of the timing capacitors, C64-C72. The fine timing is obtained by controlling the amplitude of the voltage swing at the collector of Q27 via D25. Diode D25 obtains its clamping level via emitter-follower Q30, emitter-follower Q31, and trimmer R95. Transistor Q31 compensates for the temperature coefficient of Q30 and D25.

The bias of +.65 V at the base of Q28 is obtained from D22 and R82. The bias of +.31 V at the base of Q27 is obtained from D38, R174, and R175.

The +4.5 V collector supply voltage for Q27 and Q28 is obtained from emitter-follower Q51 (on Schematic 8020-3).

The outputs of the Delay One-Shot are two complementary pulses-one at the collector of Q27 and the other at the collector of Q28.

#### 3.8 DOUBLE PULSE CIRCUITRY

The two output pulses from the Delay One-Shot drive current-switch Q32 and Q33. At the collectors of Q32 and Q33 are shorted clipping lines which provide bipolar, complementary pulses of about 4 ns width. The positive 4 ns pulses at the collectors of Q32 and Q33 occur at the beginning and end of the Delay One-Shot pulse, respectively. These pulses then pass to a gating circuit which permits either one or both pulses to pass.

If the SINGLE/DOUBLE PULSE toggle is in the SINGLE PULSE position the base of Q34 goes positive through R106. Transistor Q34 conducts and the collector of Q32 is held at nearly 0 V. This shorts out the pulse at the collector of Q32. However, the pulse at the collector of Q33 passes through diode D 29, emitter - follower Q36 and then on to trigger the base of Q39. The function of D29 is to bias on Q36 slightly. The negative excursion of the bipolar trigger pulse is below the conduction threshold and does not appear at the emitter of Q36.

If the SINGLE/DOUBLE PULSE toggle is in the Double Pulse position, the base of Q34 is at 0 V and the transistor does not conduct. The trigger pulse at the collector of Q32 then appears at the emitter of Q35 in a similar route as described in the preceding paragraphs. The two trigger pulses then appear at the base of Q39 with a time delay set by the Delay controls to provide a double pulse operation.

#### 3.9 WIDTH ONE-SHOT

The Width One-Shot, Q38-Q42, functions in an identical manner to the Delay One-Shot described above in Section 3.7. The output of the Width One-Shot is at the collector of Q38. This pulse is delayed from the TRIG OUT pulse as set by front panel Delay controls and has a width set by front panel Width controls. It is next shaped by the Output Pulse shaper for the output circuitry.

#### 3.10 OUTPUT PULSE SHAPER & OUTPUT . CIRCUITRY

Transistor Q43 and Q44 are a current-switch which receives the Width One-Shot pulse. The base of Q44 is biased at +3.9 V by divider R170, R159. The positive pulse at the base of Q43 causes a reversal of state and a negative pulse appears at the collector of Q43. This pulse, in turn, switches the output currentswitch, Q46 and Q47. A negative pulse appears at the collector of Q47 which passes to the Output connectors. Transistor Q45 is an emitterfollower to provide -2.7 V collector supply voltage for Q43 and Q44.

Transistor Q48 is a constant current supply for the output current-switch Q46 and Q47. The output amplitude is controlled by adjusting the AMPLITUDE trimmer at the base of Q48. This trimmer controls the amount of current being switched into the output load resistance.

Transistor Q49 is an emitter-follower to provide the bias level for the base of Q47. This bias level may be adjusted internally by varying R152, the Output Drive trimmer. This adjustment is adjusted to achieve the best compromise between output rise time and pulse top aberrations.

#### 3.11 OTHER CIRCUIT & CONSIDERATIONS

Emitter-follower, Q51 (Schematic 8020-3) provides the +4.5 V supply for the instrument. Note that 20 mA of current for the +4.5 V bus comes from the +24 V supply via R151. Another approximately 20 mA comes from Q51 whose main function is to regulate this bus.

Resistors R 141, R 156, and R 157 (Schematic 8020-3) are provided to nearly equalize the current from all power supply buses for optimum utilization of the NIM power supply.

#### \_ 3.12 SERVICE

If service is required on the Model 8020 refer to the schematics for the proper waveforms and voltages. A 50 MHz oscilloscope and a 350 ps sampling oscilloscope is required to properly trouble shoot this instrument.

When circuit tracing it is particularly necessary to observe the 3-4 ns wide trigger pulse has a minimum amplitude of 0.6 V at the inputs of the Delay and Width One-shots (bases of Q27 and Q28).

There is one adjustment trimmer in the Model 8020, R152 which is at the rear of the P.C. board next to the power connector. This trimmer adjusts the bias level on the output current switch. While the output pulse is observed on a sampling oscilloscope adjust the trimmer for the best compromise between the fastest output rise time and minimum pulse top aberrations. The correct setting will provide 1 ns output pulse rise time and pulse top aberrations of less than 5%.

### PARTS LIST

### **Abbreviations**

ceramic	μH	microhenry
composition carbon	μF	microfarad
electrolytic, metal case	pF	picofarad
mica	pos	positions
mylar	tan 🗉	tantalum
kilohm	v	working volts DC
megohm	var	variable
milli	• W	watts
metal film	ww	wirewound
	ceramic composition carbon electrolytic, metal case mica mylar kilohm megohm milli metal film	ceramic $\mu$ Hcomposition carbon $\mu$ Felectrolytic, metal casepFmicaposmylartankilohmVmegohmvarmilliWmetal filmww

----- NOTE -----

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The last number after each part description is the BERKELEY NUCLEONICS part number for reordering.

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### Capacitors

C1 C2 C3 C4 C5	.05 μF .05 μF 10 pF 180 pF 180 pF	cer cer cer cer cer	100 V 100 V 1 kV 1 kV 1 kV 1 kV		110-017 110-017 110-002 110-007 110-007	C37 C38 C39 C40 C41	$\begin{array}{c} .0068 \ \mu F \\ .022 \ \mu F \\ .068 \ \mu F \\ .22 \ \mu F \\ .68 \ \mu F \end{array}$	myl myl myl tan tan	600 V 100 V 100 V 35 V 35 V	10% 10% 10% 10%	114-009 114-014 114-019 122-007 122-009
C6 C7 C8 C9 C10	180 pF 180 pF 5 pF 180 pF 5 pF	cer cer cer cer cer	1 kV 1 kV 1 kV 1 kV 1 kV 1 kV		110-007 110-007 110-001 110-007 110-001	C43 C44 C45 C46 C47	.05 μF 10 pF 180 pF 180 pF 180 pF 180 pF	cer cer cer cer cer	100 V 1 kV 1 kV 1 kV 1 kV 1 kV		110-017 110-002 110-007 110-007 110-007
C11 C12 C13 C14 C15	.05 μF 180 pF 180 pF 180 pF 5 pF	cer cer cer cer cer	100 V 1 kV 1 kV 1 kV 1 kV 1 kV		110-017 110-007 110-007 110-007 110-001	C48 C49 C50 C51 C52	180 pF 180 pF 10 pF 180 pF 180 pF	cer cer cer cer cer	1 kV 1 kV 1 kV 1 kV 1 kV 1 kV	. '	110-007 110-007 110-002 110-007 110-007
C16 C17 C18 C19 C20	180 pF 100 pF 100 pF 2.5-10 pF 180 pF	cer cer cer var cer	1 kV 1 kV 1 kV 1 kV		110-007 110-006 110-006 130-004 110-007	C53 C54 C55 C56 C57	5 pF 180 pF 180 pF 180 pF 180 pF	cer cer cer cer cer	1 kV 1 kV 1 kV 1 kV 1 kV		110-001 110-007 110-007 110-007 110-007
C21 C22 C23 C24 C25	180 pF 180 pF 180 pF 180 pF 180 pF 180 pF	cer cer cer cer cer	1 kV 1 kV 1 kV 1 kV 1 kV 1 kV		110-007 110-007 110-007 110-007 110-007	C58 C59 C60 C61 C62	180 pF 180 pF 180 pF 25 pF 10 pF	cer cer cer cer mic	1 kV 1 kV 1 kV 1 kV 1 kV 500 V	5%	110-007 110-007 110-007 110-003 112-016
C26 C27 C28 C29 C30	180 pF 180 pF 25 μF 180 pF 180 pF	cer cer elec cer cer	1 kV 1 kV 25 V 1 kV 1 kV		110-007 110-007 120-005 110-007 110-007	C63 C64 C65 C66 C87	180 pF .33 μF .1 μF .033 μF .01 μF	cer tan myl myl myl	1 kV 35 V 100 V 100 V 400 V	10% 10% 10% 10%	110-007 122-008 114-020 114-024 114-011
C31 C33 C34 C35 C36	180 pF 22 pF 100 pF 430 pF .002 μF	cer mic mic mic myl	1 kV 500 V 500 V 100 V 600 V	5% 5% 5% 10%	110-007 112-001 112-004 112-012 114-002	C68 C69 C70 C71 C72	.0033 μF .001 μF 270 pF 68 pF 18 pF	myl myl mic mic mic	600 V 600 V 500 V 500 V 500 V	10% 10% 5% 5%	114-005 114-001 112-009 112-018 112-017

continued

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JAFA	CITORS (co	ontinued)			Diod	es				
C75	180 pF	cer	1 kV	110-007	D1		1N415	4		411-003
C76	180 pF	cer	1 kV	110-007	D2		1N415	4		411-003
C77	10 pF	cer	1 kV	110-002	D3		1N415	4		411-003
C78	180 pF	cer	1 kV	110-007	D4		1N415	4		411-003
C79	180 pF	cer	1 kV	110-007	D5		1N415	4		411-003
C80	180 nF	cer	1 kV	110-007	D6		1N415	4		411-003
C81	180 pr	Cer		110-007	D7		FH110	0		415-001
Č82	10  pr	cer	1 kV	110-002	D8		1N415	4		411-003
C83	180 pF	cer	Î kV	110-007	D9		1N415	4		411-003
Č84	180 pF	cer	1 kV	110-007	DIO		1N415	4		411-003
					D11		1N415	4		411-003
CSD	10 pr_	mic	500 V 5%	b 112-010	DIZ		FMELC	10		415-001
C86	.33 µr	tan	30 V 10	% 122-008	D13		1 111	<i>A</i>		410-001
000	·ιμr	myi	100 V 10	0 114-020	D14 D15		111410	1		411-003
Č89	.033 μF .01 μF	myl	400 V 10	% 114-011	Dig		11/210	<b>T</b>		411-000
	•	•			D16		1N415	4		411-003
C90	.0033 JIF	mvl	600 V 10	% 114-005	D17		1N415	4		411-003
Č91	.001 UF	mvl	600 V 10	<b>%</b> 114-001	D18		1N415	4		411-003
Č92	270 pF	mic	500 V 59	6 112-009	D19		1N410	4		411-003
C93	68 pF	mic	500 V 59	6 112-018	1)20	1	11410	4		411-003
C94	10 pF	mic	500 V 59	6 112-016	D21		1N415	4		411-003
					D21		FH110	้าดิ		415-001
C96	180 pF	cer	1 kV	110-007	<b>D23</b>		1N415	4		411-003
Č99	180 pF	cer	1 kV	110-007	D24		1N415	4		411-003
C100	180 pF	cer	1 kV	110-007	D25		FH11(	)0		415-001
C101	180 pF	cer	1 kV	110-007						
C102	180 pF	cer	1 kV	110-007	D26		1N415	4		411-003
					D28		1N415	4		411-003
C103	5 pF	cer	1 kV	110-001	D29		1N415	94		411-003
C104	.05 μF	cer	100 V	110-017	D30		FHII(	30		410-001
C105	180 pF	cer	1 kV	110-007	D91		111410	14		411-003
C106	180 pF	cer	1 kV	110-007	D32		1 N 4 1 5	4		411-003
C107	$.05 \ \mu F$	cer	100 V	110-017	D32		FH11	50		415-001
					D34		1 N41	54		411-003
					D35		1N415	54		411-003
C108	20 pF_	cer	1 kV	110-018	$\tilde{D}37$		1N415	54		411-003
C109	100 µF	elec	10 V 1 1-37	120-007						
CHU	180 pr	cer	1 67	110-007	D40		1N <b>52</b> 3	81		412-009
0120	100 pr 50	aloa	95 V	120_008	D41		FH11	00		415-001
CITE	50 μΡ	erec	4 <b>5</b> Y	120-000	D42		FH11	00		415-001
C119	50	مام	95 V	120-006	Indu	ctore				
C114	50 μF	elec	25 V	120-006	mao	ciors				
C115	50 µ.F	elec	25 V	120-006	L1		.33 µ	H		310-011
C116	180 pF	cer	1 kV	110-007	L2		.22 μ	H		310-013
C117	180 pF	cer	1 kV	110-007	L3		<b>. 22</b> μ	H		310-013
C118	180 pF	cer	1 kV	110-007	Resi	stors				
C119	180 pF	cer	1 kV	110-007					- 01	010 <b>/</b> 8 /
C120	10 pF	cer	1 kV	110-002	R1	470 k	1/4 W	comp	5%	213-474
C121	$.05 \ \mu F$	cer	100 V	110-017	R2	270 k	1/4 W	comp	5% 50	213-274
C122	.05 µF	cer	100 V	110-017	R3	10	1/4 W	comp	270	213-100
				•	H4	I K	$\frac{1}{4} W$	comp	570	213-104
~100	05		100 37	110 017	Rə	91	1/4 W	comp	<i>U</i> /0	210-010
C123	.05 μF	Cer	100 V 1 bV	110-017	<b>P</b> 6	51	1/4 W	comp	5%	213-510
C124	100 pr	Cer	100 17	110-017	R7	51	1/4 W	comp	5%	213-510
C120	05 µr	Cer	100 V	110-017	R8	1 k	1/4 ŵ	comp	5%	213-102
C127	180 nF	Cer	1 kV	110-007	R9	8.2 k	1/4 W	comp	5%	213-822
0141	100 pr		2		R10	56	1/4 W	comp	5%	213-560
C120	05	¢er	100 V	110-017	R11	56	1/4 W	comp	5%	213-560
C120	180 pF	Cer	1 kV	110-007	<b>R12</b>	270	1/4 W	comp	5%	213-271
C130	10 pF	cer	1 kV	110-002	RIS	10 k 👘	1/4 W	comp	5%	213-103
C131	10 pF	cer	1 kV	110-002	<b>R14</b>	240	1/4 W	comp	5%	213-241
Č132	180 pF	cer	1 kV	110-007	R15	51	∽ 1/4 W	comp	5%	213-510

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R16 R17 R18 R19 R20	22 k 15 k 33 51 51	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	558888 555555	213-223 213-153 213-330 213-510 213-510 213-510	R76 R77 R79 R80 R81	510 150 51 560 56	1/4 W 1/4 W 1/4 W 1/2 W 1/2 W	comp comp comp comp	5%%%%%% 55%%%%%%%%%%%%%%%%%%%%%%%%%%%%	213-510 213-151 213-510 212-561 213-560
R21 R22 R23 R24 R25	51 1.2 k 8.2 k 27 360	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	55558 55555555555555555555555555555555	213-510 213-122 213-822 213-270 213-270 213-361	R82 R83 R84 R85 R86	2.7 k 4.7 k 5.6 k 10 k 560	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	5% 5% 5% 5% 5%	213-272 213-472 213-562 213-103 213-561
R26 R27 R28 R29 R30	560 8.2 k 1.8 k 270 110	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	5%%%%% 5%%%%% 55%	213-561 213-822 213-182 213-271 213-271 213-111	R87 R88 R89 R90 R91	1.2 k 1 k 390 91 51	1/2 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	5%%%%% 5%%%%%% 55%	212-122 213-102 213-391 213-910 213-510
R31 R32 R33 R34 R35	1.5 k 6.8 k 150 150 750	1/2 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	55888888 55555555555555555555555555555	212-152 213-682 213-151 213-151 213-151 213-751	R92 R93 R94 R95 R96	4.7 k 2.2 k 5.6 k 1 k 1 k	1/4 W 1/4 W 1/4 W 1 W 1/4 W	comp comp comp trimmer comp	5% 5% 5%	213-472 213-222 213-562 244-017 213-102
R36 R37 R38 R39 R40	750 39 39 470 6.8 k	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	55555555	213-751 213-390 213-390 213-471 213-682	R98 R102 R103 R104 R105	470 470 51 91 100	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	5%%%%% 5%%%%% 5%%	213-471 213-471 213-510 213-910 213-101
R41 R42 R43 R44 R45	220 10 k 1 k 4.7 k 22	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	5%%%% 55%%% 55%	213-221 213-103 213-102 213-472 213-220	R106 R107 R108 R109 R110	10 k 4.7 k 4.7 k 51 56	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	55555555	213-103 213-472 213-472 213-510 213-560
R46 R47 R48 R49 R50	22 390 470 10 k 15 k	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	5%%%%% 55% 55%	213-220 213-391 213-471 213-103 213-153	R111 R112 R113 R114 R115	100 33 560 56 2.7 k	1/4 W 1/4 W 1/2 W 1/4 W 1/4 W	comp comp comp comp	5%%%% 55%%%	213-101 213-330 212-561 213-560 213-272
R51 R52 R53 R54 R55	4.7 k 4.7 k 1 k 1.2 k 39 k	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	5 5 5 5 5 5 5 5 5 5 5	213-472 213-472 213-102 213-122 213-393	R116 R117 R118 R119 R120	15 k 10 k 560 5.6 k 51	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	55% 55% 55%	213-153 213-103 213-561 213-562 213-510
R56 R57 R58 R59 R60	2.2 k 1.2 k 1 k 820 1.6 k	1/4 W 1/4 W 1 W 1/4 W 1/4 W	comp comp trimme: comp comp	5% 5% 5% 5%	213-222 213-122 244-017 213-821 213-162	R121 R122 R123 R124 R125	1.2 k 4.7 k 2.2 k 22 k 1 k	1/2 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	5%%%%% 55% 55 55 5	212-122 213-472 213-222 213-223 213-102
R61 R62 R63 R64 R65	10 k 910 51 51 51	1/4 W 1/2 W 1/4 W 1/4 W 1/4 W	comp comp comp comp comp	5%%%% 5%%%% 5%	213-103 212-911 213-510 213-510 213-510 213-510	R126 R127 R132 R133 R134	1 k 5.6 k 3.3 k 7.5 k 100	1 W 1/4 W 1/4 W 1/4 W 1/4 W	trimme: comp comp comp comp	r 5%% 5%% 5%	244-017 213-562 213-332 213-752 213-101
R66 R67 R68 R69 R70	510 100 4,7 k 1 k 470	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	5%%%% 5%%%% 55%	213-511 213-101 213-472 213-102 213-471	R136 R137 R138 R139 R141	33 51 51 330 180	1/4 W 1/4 W 1/4 W 1/2 W 2 W	comp comp comp comp	5% 5% 5% 5% 5%	213-330 213-510 213-510 212-331 212-331 210-181
R71 R72 R73 R74 R75	15 k 6.8 k 51 51 51	1/4 W 1/4 W 1/4 W 1/4 W 1/4 W	comp comp comp comp	5%%%% 5%%%% 55%%	213-153 213-682 213-510 213-510 213-510 213-510	R142 R143 R144 R145 R146	680 1 k 130 3.3 k 2.4 k	1/4 W 1 W 1/4 W 1/4 W 1/4 W	comp trimme: comp comp comp	5% r 5% 5%	213-681 244-017 213-131 213-332 213-242

**RESISTORS** (continued)

RESIST	rors (c	ontinued)				Q16 017	MPS2924 MPS6531	430-009
R147	470	1 W	comp	5%	211-471	Q18	MPS3638	430-010
R148	39	1/4 W	comp	5%	213-390	Q19	2N5179	430-025
R149	91	1/4 W	comp	5%	213-910	Q20	2N5179	430-025
R150	91	1/4 W	comp	5%	213-910			
R191	750	1 W	comp	5%	211-751	Q21	2N5179	430-025
D159	10 1-	1 337	tanà amin'ny same		944-019	· Q22	2N5179	430-025
R159	15 6	1/4 W	comp	502	213-153	Q23	MPS6534	430-018
R154	470	1/4 W	comp	5%	213-471	Q24	2N4258	430-030
R155	1.8 k	1/4 w	comp	5%	213-182	Q25	2N4258	430-030
R156	180	$\bar{2}'\bar{\mathbf{W}}''$	comp	5%	210-181	1		
		1 A 14				Q26	2N5179	430-025
R157	180	2 W	comp	5%	210-181	Q27	2N5179	430-025
R158	33	1/4 W	comp	5%	213-330	Q28	2N5179	430-025
R159	4,7 k	1/4 W	comp	5%	213-472	Q29	2N5179	430-025
R160	33	1/4 W	comp	5%	213-330	- <b>Q</b> 30	MP\$6531	430-017
RTOT	1.0 K	1/4 W	comp	970	213-102			
D162	10 1	1/4 32/	comp	596	213-103	Q31	MPS3638	430-010
R168	51	1/4 W	comp	5%	213-510	Q32	2N4258	430-030
<b>R169</b>	51	1/4 W	comp	5%	213-510	Q33	2N4258	430-030
R170	10	1/4 W	comp	5%	213-100	Q34	2N5179	430-025
R171	51	1/4 W	comp	5%	213-510	କ୍ଷର୍ବର	2N2179	430-020
		•	. ·		· · · · ·		· · · · ·	
R172	. 10	1/4 W	comp	5%	213-100	Q36	2N5179	430-025
R173	2.7 k	1/4 W	comp	5%	213-272	Q37	2N5179	430-025
R174	33	1/4 W	comp	370	213-330	୍ୟୁଅଷ	2N0179	430-025
TC170	3.9K	1/4 W	comp	270	213-382	 	MP30034 9N5170	430-018
R177	3 9 k	1/4 W	comp	59	213-392	6.20	2140110	-30-020
R178	120	1/4 W	comp	5%	213-121			
		-/		• /0		Q41	MPS6531	430-017
<b>.</b>		·		- <sup>1</sup>	·	Q42	MP83038	430-010
Trans	sistors			$\mathcal{L}_{i} = \mathcal{L}_{i}$		044	2N 4757 9N 4050	490-022
· ·		0.55	~		100 000	045	MD88534	430-018
Ω.		2N017	A.		430-025	<b>4</b> ,10	#11 D0001	100-010
<b>W4</b>		4N017 9N16177	9 )		430-020	0.40	01-54.0.0	400.004
04 04		2N017	0		430-045	Q40	2N5109	430-024
<b>0</b> 5	1. S.	2N517	0 0		430-025	Q91	2N0109 9N9910	490.004
						040	MDR6531	430-017
Q6		MPS2	924		430-009	650	MP86531	430-017
Q7		2N517	9		430-025			100-011
Q8		2N517	9		430-025	Q51	MP86531	430-017
Q9		2N517	9		430-025	Q52	MPS6534	430-018
Q10		2N517	9		430-025	Q53	2N5179	430-025
011		01754 8	•		400 00F	Q54	MPS6518	430-016
		2N017	0 A		430-020	<b>Q</b> 55	MPS6518	430-016
019		ADDA1	87 591		100-040 17			400 010
014		2N517	0.		490-011	<b>W</b> 00	MPS0018	430-016
Å15		211517	Ő.		430-025	401 050	MP30010	430-010
-46 T V		DIANT (	v		100-040	M00	MP06764	490-008

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